



Cyprus
University of
Technology

EEN320 - Power Systems I (Συστήματα Ισχύος Ι)

Part 8: The transmission line

<https://sps.cut.ac.cy/courses/een320/>

Dr Petros Aristidou

Department of Electrical Engineering, Computer Engineering & Informatics

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After this part of the lecture and additional reading, you should be able to . . .

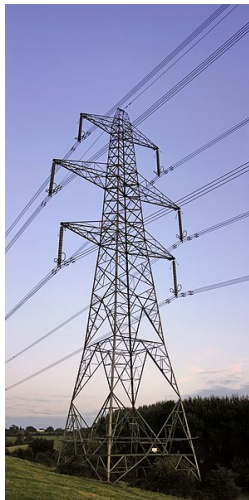
- 1 . . . name the main components of an overhead line and describe their functionality;
- 2 . . . understand how to derive the Π -equivalent circuit of a transmission line;
- 3 . . . determine the parameters of the Π -equivalent circuit of a transmission line from its concentrated parameters;
- 4 . . . explain how and under which conditions or assumptions the Π -equivalent circuit can be further simplified.

Outline

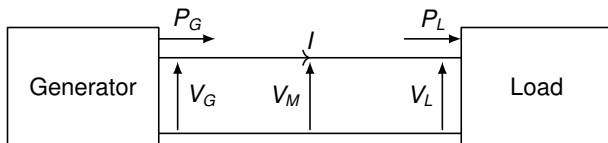
- 1 Physical relevance of power lines**
- 2 Structure of overhead lines**
 - Conductors
 - Support structures
 - Insulators
 - Shield wires
- 3 Derivation of lumped inductor and capacitor values**
 - Inductance
 - Capacitance
- 4 Overhead line parameters**
 - Concentrated parameters
 - Some brief remarks on cables
- 5 Equivalent circuits for power lines**
 - Differential equation of a power line
 - Solution of differential equation of a power line
 - Π -equivalent circuit
 - Model simplifications and their validity

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- Task: Transport electricity
- 2 main types:
 - Overhead line (OL)
 - Cable
- OLs and cables possess different structure and operational properties
- At same voltage level, costs for cables approx. 10-20 times than costs of OLs
- OLs economically more viable option



Overhead power line in Gloucestershire, England ©Yummifruitbat



Simplified DC transmission system

- Line is resistive \rightarrow voltage drop across line $\rightarrow V_G > V_M > V_L$
- Average power transmitted over line: $P_{trans} = V_M I$
(V_M is voltage at middle of line length)
- Denote total line resistance by $R \rightarrow$ line losses given by

$$P_{loss} = RI^2 = R \left(\frac{P_{trans}}{V_M} \right)^2$$

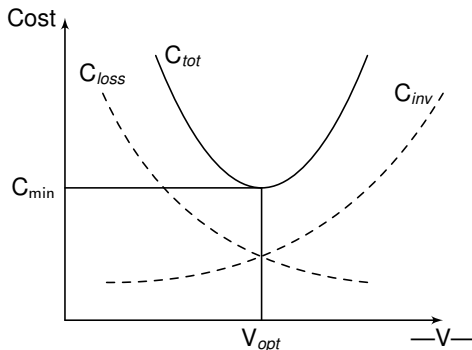
- Ratio of power losses to transmitted power

$$\frac{P_{loss}}{P_{trans}} = R \frac{P_{trans}}{V_M^2}$$

- Ratio of power losses to transmitted power

$$\frac{P_{loss}}{P_{trans}} = R \frac{P_{trans}}{V_M^2}$$

- Power losses inversely proportional to square of operational voltage V_M^2
 - Power lines usually operated at high voltage
 - However, higher voltage means higher insulation of components
- Higher costs

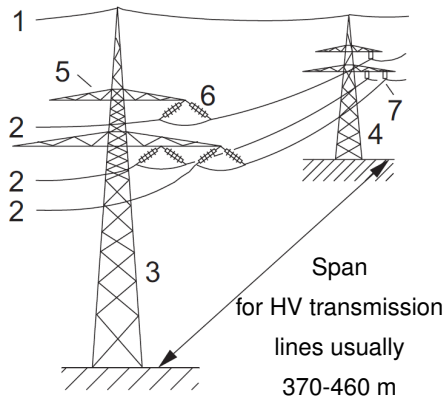


- Total costs $C_{tot} = C_{loss} + C_{inv}$
- Minimum costs $C_{min} \rightarrow$ economically optimal operating voltage

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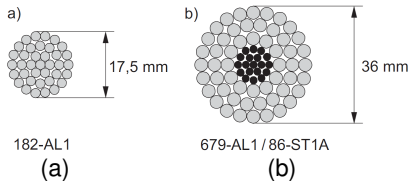
An overhead line consists of

- Conductors (2)
- Support structures
 - Towers (and poles) (3,4)
 - Traverse (5)
- Shield wires (1)
- Insulators (6,7)
 - Strain-type insulator (6)
 - Suspension-type insulator (7)



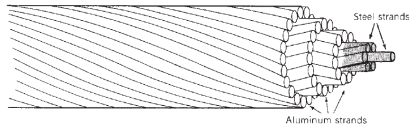
Source: K. Heuck et al., "Elektrische Energieversorgung", Springer, 2013

- Aluminium is most common conductor metal
- Copper also used, but less frequently as heavier and more expensive
- Mechanical strain acting on conductors limits span between towers (line sag)
- For short lines, conductors of pure aluminium strands may be used (Figure (a))
- For longer lines, conductors are *reinforced* with central steel strands (Figure (b))



Source: K. Heuck et al., "Elektrische Energieversorgung", Springer, 2013

Strands are usually twisted to reduce Eddy currents

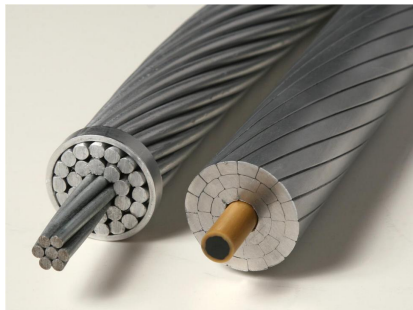


Source: J. Duncan Glover et al., "Power System Analysis & Design", Cengage Learning, 2008

- Common types of conductors
 - Aluminium conductor steel-reinforced (ACSR)
 - All-Aluminium conductor (AAC)
 - All-Aluminium alloy conductor (AAAC)
 - Aluminium conductor composite reinforced (ACCR)
 - Aluminium conductor composite core (ACCC)
- Conductors labeled based on cross section (in mm^2) of aluminium and core strand

Example: 243-AL1/39-ST1A

(code after AL and ST denotes finishing properties of AL and ST)



©Dave Bryant

Standard round-wire ACSR (left) and ACCC with trapezoidal wires (right)

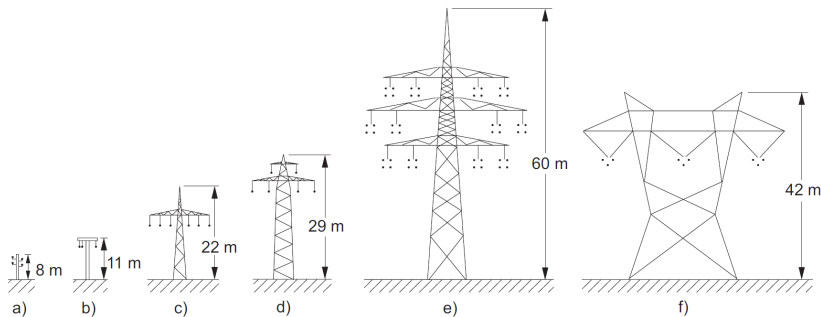
ACCR and ACCC use carbon and glass fiber core → up to 10 times lower thermal expansion coefficient than steel → can use more aluminium → reduced line losses

- Each phase of a three-phase transmission line consists of one or more conductors
- More than one conductor/phase → bundled conductor
- Advantages:
 - Smaller series resistance
 - Reduced electric field strength at conductor surface → reduced Corona effect
- Transmission line may also consist of several three-phase systems in parallel



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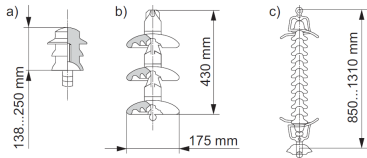
Triple-circuit 400 kV overhead line with
4 conductors per phase



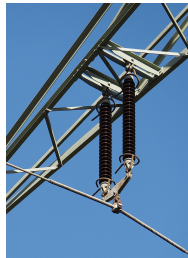
Source: K. Heuck et al., "Elektrische Energieversorgung", Springer, 2013

- Large variety of support structures
- Poles made of wood or concrete [a) and b)] used for voltages ≤ 110 kV
- Self-supporting lattice steel towers [c) - f)] used for voltages ≥ 110 kV

- Need to insulate "live" conductors from tower
- Pin-type insulators (for lower voltages < 60 kV); material: porcelain; Figure (a)
- Suspension-type insulators (for voltages > 60 kV)
 - Suspension disc insulator; material: glas; Figure (b)
 - Long-rod insulator; material: porcelain; Figure (c)
(Strain-type insulator some times also used)
- To prevent sparkovers, insulators need to be sufficiently long (approx. 1.5 cm/kV) and possess appropriate shape to minimise leakage currents

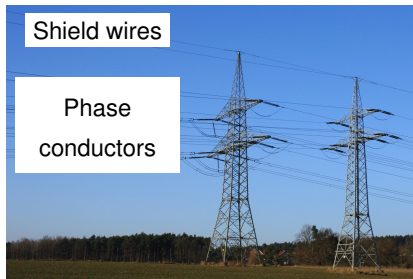


Source: K. Heuck et al., "Elektrische Energieversorgung", Springer, 2013



110 kV double long-rod suspension string
©Kreuzschnabel

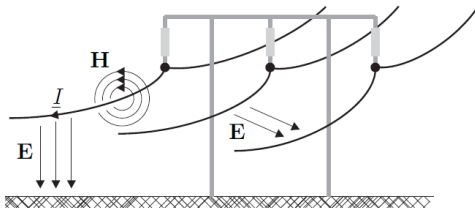
- Shield wires located above phase conductors to provide protection against lightning
- Shield wires are grounded to tower
- They also serves as parallel path with Earth for fault currents
- Predominantly used above 110 kV
- Much smaller cross section than phase conductors
- Modern shield wires contain optical fibres for communication/control
- Usually, 1-2 shield wires used



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Magnetic and electric fields of conducting power line



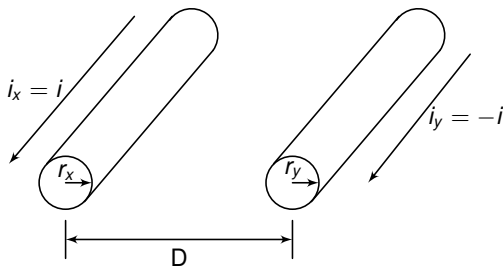
E... electric field

H... magnetic field

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- Due to the alternating current, there is an alternating magnetic field in each line affecting neighbouring lines.
- Similarly, there is an electric field between lines and from lines to the ground.
- The parameters of the line model are dictated by the line material (ohmic losses) as well as the magnetic and electric fields.

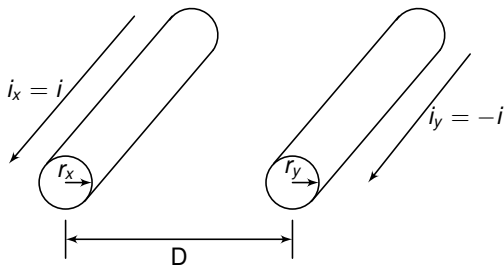
3.1 Inductance of a single-phase two-wire line (1)



- r_x, r_y : radius of cylindrical conductors
- D : spacing between conductors
- i : current flowing in conductors
- **Assumptions:** Conductors are of infinitely length, non-magnetic ($\mu = \mu_0$)¹ and have uniform current density (skin-effect neglected)

¹ μ_0 is vacuum permeability constant

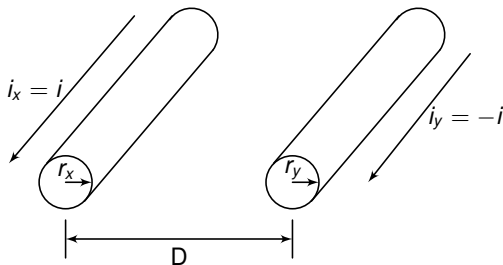
3.1 Inductance of a single-phase two-wire line (2)



- Inductance of one conductor $k = \{x, y\}$

$$L'_k = \frac{\mu_0}{2\pi} \ln \left(\frac{D}{r'_k} \right) = 2 \cdot 10^{-7} \ln \left(\frac{D}{r'_k} \right) \text{ [H/m]} \quad r'_k = r_k e^{-\frac{1}{4}} \approx 0.778r$$

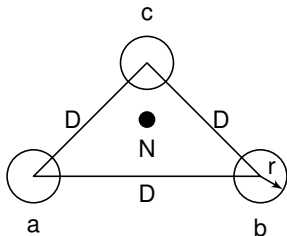
$$\mu_0 = 4\pi \cdot 10^{-7}$$



- Total inductance of single-phase two-wire line

$$\begin{aligned}
 L' &= L'_x + L'_y = 2 \cdot 10^{-7} \left(\ln \left(\frac{D}{r'_x} \right) + \ln \left(\frac{D}{r'_y} \right) \right) \\
 &= 2 \cdot 10^{-7} \ln \left(\frac{D^2}{r'_x r'_y} \right) = 4 \cdot 10^{-7} \ln \left(\frac{D}{\sqrt{r'_x r'_y}} \right) \text{ [H/m]}
 \end{aligned}$$

- Identical conductors ($r_x = r_y$): $L' = 4 \cdot 10^{-7} \ln \left(\frac{D}{r'} \right)$ [H/m]



- Assumptions: balanced phase currents, equidistant spacing D , identical conductor radii r
- Line-neutral inductances of three-phase three-wire line

$$L' = L'_a = L'_b = L'_c = 2 \cdot 10^{-7} \ln \left(\frac{D}{r'} \right) \text{ [H/m]}$$

- This is half the inductance of a single-phase two-wire line!
- Inductances balanced \rightarrow can use single-phase equivalent circuit for network calculations

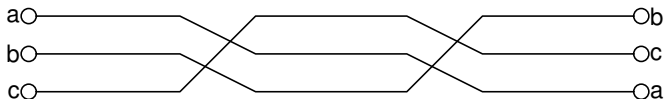
3.1 Inductance of a three-phase three-wire line - Transposition of conductors

- In practice, conductors rarely spaced in equidistant manner
- Inductances become unbalanced ($L_a \neq L_b \neq L_c$) → this causes unbalanced voltage drops even if currents are balanced!
- Practical remedy: restore balance by exchanging conductor positions along line (e.g. at substations)
- This is called *transposition*
- For transposed line with equivalent spacing $D_{eq} = \sqrt[3]{D_{12}D_{23}D_{31}}$

$$L' = L'_a = L'_b = L'_c = 2 \cdot 10^{-7} \ln \left(\frac{D_{eq}}{D_S} \right) \text{ [H/m]}$$

D_S ... Geometric Mean Radius (GMR) for stranded conductors

$D_S = r'$ for solid cylindrical conductors



3.1 Example: Determine inductance of a three-phase three-wire line

Task. A completely transposed 50-Hz three-phase line has flat horizontal phase spacing with 10m between adjacent conductors. The geometric mean radius (GMR) of the conductors is 0.0159m. The line length is $\ell = 200\text{km}$. Determine the inductance in H and the inductive reactance in Ω .

Solution. We have that

$$D_{eq} = \sqrt[3]{10 \cdot 10 \cdot 20} = 12.6 \text{ m}$$

Hence,

$$L' = 2 \cdot 10^{-7} \ln \left(\frac{D_{eq}}{D_S} \right) = 2 \cdot 10^{-7} \ln \left(\frac{12.6}{0.0159} \right) = 1.335 \mu \text{ H/km}$$

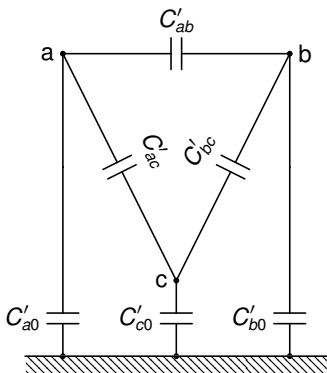
and

$$L = L' \cdot \ell = 1.335 \cdot 10^{-6} \cdot 200 = 0.267 \text{ H},$$

as well as

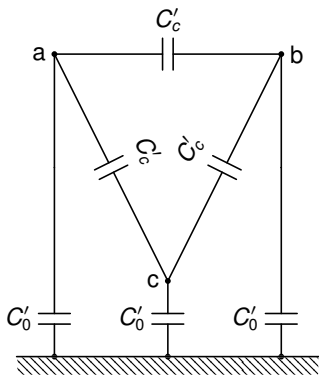
$$X = L\omega = 0.267 \cdot 2 \cdot \pi \cdot 50 = 83.88 \Omega$$

- Line capacitance can be obtained in similar fashion to inductances
- Need to consider interaction of electric fields between conductors *and* between individual conductors and earth
- This can be modelled via coupling and earth capacitances



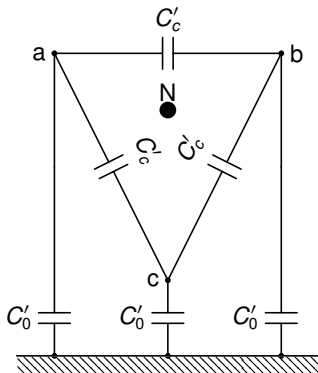
3.2 Capacitance - Balanced three-phase three-wire line (1)

- Assume balanced line (e.g. via transposition)
- Then, $C'_0 = C'_{a0} = C'_{b0} = C'_{c0}$ and $C'_c = C'_{ab} = C'_{ac} = C'_{bc}$
- Coupling conductors C'_c form Δ -connection



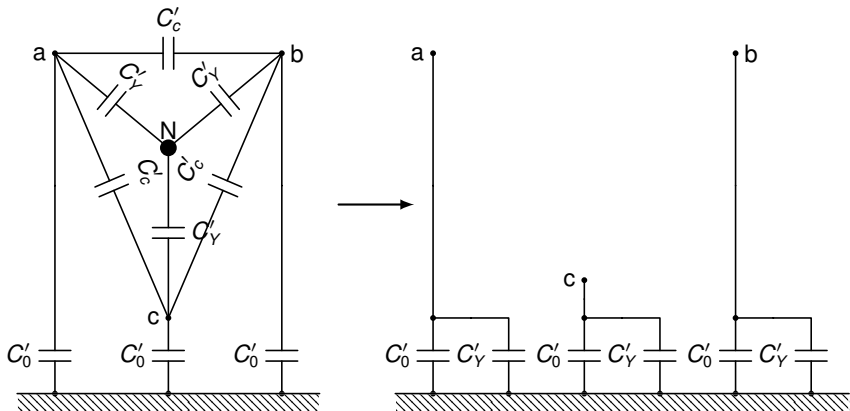
3.2 Capacitance - Balanced three-phase three-wire line (2)

- Assume balanced line (e.g. via transposition)
- Then, $C'_0 = C'_{a0} = C'_{b0} = C'_{c0}$ and $C'_c = C'_{ab} = C'_{ac} = C'_{bc}$
- Coupling conductors C'_c form Δ -connection
- Introduce *fictitious* neutral point N



3.2 Capacitance - Balanced three-phase three-wire line (3)

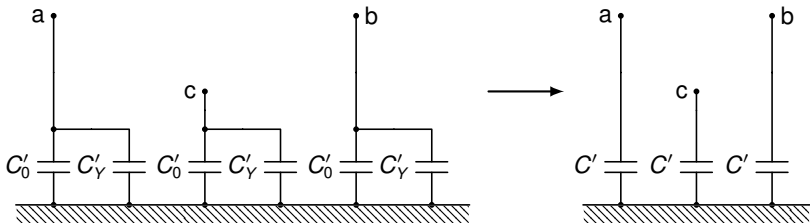
- Balanced conditions \rightarrow sum of currents at N equal zero \rightarrow N has same potential as ground
- Parallel connection of coupling and earth capacitances $C'_Y = 3C'_c$



- Balanced conditions \rightarrow sum of currents at N equal zero \rightarrow N has same potential as ground
- Parallel connection of coupling and earth capacitances $C' = C'_Y + C'_0$

$$C' = \frac{2\pi\epsilon_0}{\ln\left(\frac{D_{eq}}{r}\right)} \text{ [F/m]} \quad \epsilon_0 \dots \text{vacuum permittivity}$$

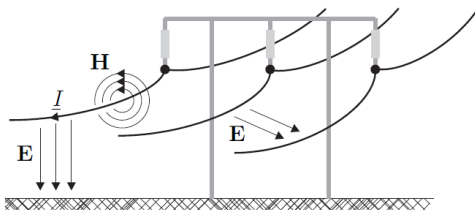
- Typical value for overhead lines $C' \approx 10 \text{ nF/km}$



Note: similar calculations applicable to conductor bundles

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Magnetic and electric fields of conducting power line



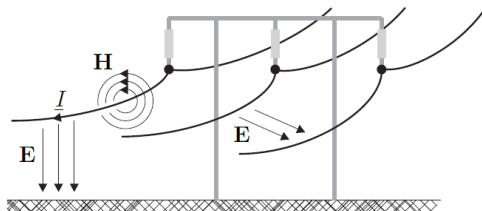
E... electric field

H... magnetic field

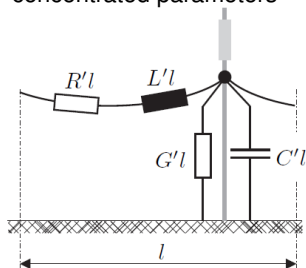
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- Each power line has characteristic line parameters
- Parameters dependent on line geometry and material
- Parameters often indicated in [unit]/km and by giving the line length ℓ

Magnetic and electric fields
of conducting power line



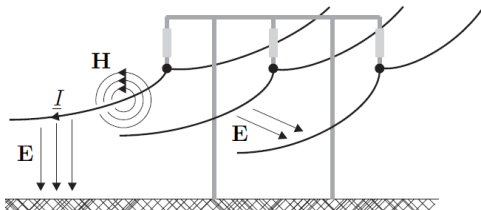
Power line model with
concentrated parameters



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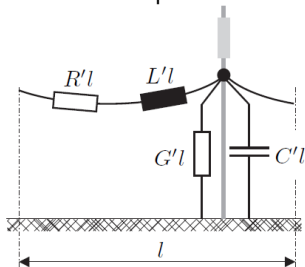
- Line resistance R' [Ω/km] \leftrightarrow Ohmic resistance of conductor
- Line inductance L' [H/km] \leftrightarrow Magnetic field of conductor
- Capacitance C' [F/km] \leftrightarrow Electric field of conductor
- Shunt conductance G' [S/km] \leftrightarrow Leakage currents at insulators

Magnetic and electric fields
of conducting power line



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Power line model with
concentrated parameters



- For performing circuit analysis involving power lines (e.g. to determine the network conditions or design) we need to know the concentrated parameters of the lines
- Usually, concentrated parameters indicated by manufacturer

Please see the course book for a detailed derivation.

- Real conductors are not lossless!
- This can be accounted for by including a series resistance in the conductor model
- For DC current, resistance of conductor can easily be determined from its diameter, length and specific conductivity
- For AC current, in addition the *skin effect* needs to be considered when determining the resistance of a conductor
- Skin-effect: current not distributed homogeneously over conductor diameter, but concentrated towards conductor boundaries
- Current density increases towards conductor boundaries
- Effective diameter of conductor is reduced and, hence, ohmic resistance is increased compared to DC resistance (typically by a few percent)

4.1 Resistance - Calculation

- For steel-reinforced aluminium conductors (ACSR), AC resistance is approximately same as DC resistance
- Reason: Skin-effect → reduced AC current in steel strands → increase in AC resistance by skin-effect comparable to higher DC current in steel strands
- Conductor losses result in heat dissipation → maximum conductor current limited, as long-term high temperatures ($> 80^{\circ}$) decrease mechanical strength of conductor material → line sags
- Line resistance operating at temperature of ϑ° can be calculated via

$$R' = R'_{20}(1 + \alpha(\vartheta - 20^{\circ}\text{C})) \text{ [R/m]}$$

$$R'_{20} = \frac{\rho_{20}}{A} \text{ resistance of conductor at } 20^{\circ}\text{C}$$

ρ_{20} . . . specific resistance of of conductor material at 20°C

A . . . effective conductor area

- For practical conductors, resistance values obtained via measurements

- Also, losses due to insulator leakage currents and corona
- Corona: high value of electric field strength at conductor surface causes air to become electrically ionised and to conduct
- Corona losses dependent on meteorological conditions (rain; humidity) and conductor surface irregularities
- For overhead lines, conductance G' can only be estimated from measurements, while it can be determined experimentally for cables
- Usually, conductance is very small and therefore most often neglected in power system studies

- Cables mostly used at low voltage levels (<110 kV)
- Often installed underground
- Physical characteristics of cables fundamentally different from overhead transmission lines (OHLs)!
- Main reasons:
 - Distance between conductors as well as between conductors and earth much smaller in cables than in OHLs
 - Conductors in cables typically surrounded by other metallic materials, e.g. skin
 - Insulation material of OHLs is air, while in cables materials such as paper, oil or SF₆ are used
- Consequences:
 - Inductance of OHLs usually higher as that of cables
 - Capacitance of cables usually much higher as that of OHLs

- Typical values for parameters of OHLs at 50 Hz

Rated voltage in kV	230	345	500	765
R' [Ω/km]	0.050	0.037	0.028	0.012
$X'_L = \omega L'$ [Ω/km]	0.407	0.306	0.271	0.274
$Y'_C = \omega C'$ [$\mu\text{S}/\text{km}$]	2.764	3.765	4.333	4.148

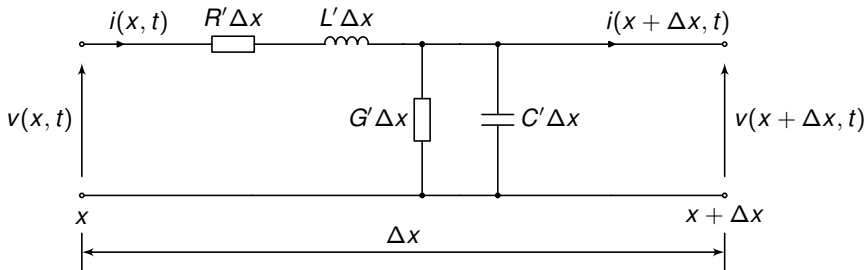
- Typical values for parameters of cables at 50 Hz

Rated voltage in kV	115	230	500
R' [Ω/km]	0.059	0.028	0.013
$X'_L = \omega L'$ [Ω/km]	0.252	0.282	0.205
$Y'_C = \omega C'$ [$\mu\text{S}/\text{km}$]	192.0	204.7	80.4

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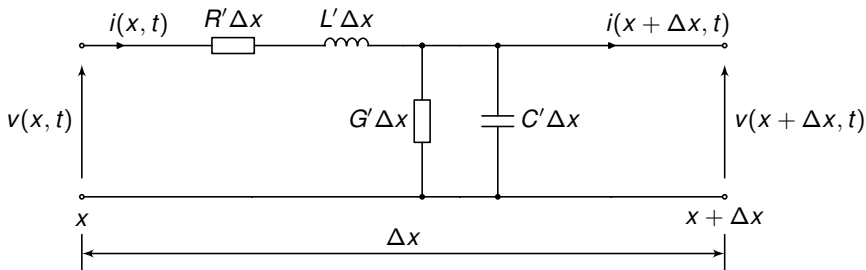
- Being able to describe the behaviour of a power systems by a mathematical model is a fundamental prerequisite for network planning and operation
 - We will derive a model of a power line that is valid under *stationary* (or steady-state) conditions
 - Note: A real power system is never exactly in steady-state due to continuous variations of load and generation
 - However, under normal conditions this variations are of small magnitude compared to overall power flows in network
 - Also, normal load patterns change over fairly long period (several tens of minutes)
- Steady-state model suitable for describing nominal network operating conditions

Section of length Δx of homogeneous power line



- Line parameters R' , L' , G' and C' are not lumped, but (uniformly) distributed along length of line; Δx denotes a small distance
- Propagation of current $i(t, x)$ and voltage $v(t, x)$ across that line segment is not instantaneous
- Propagation can be described by a partial-differential equation (i.e. propagation depends on time t and location x)

Section of length Δx of homogeneous power line



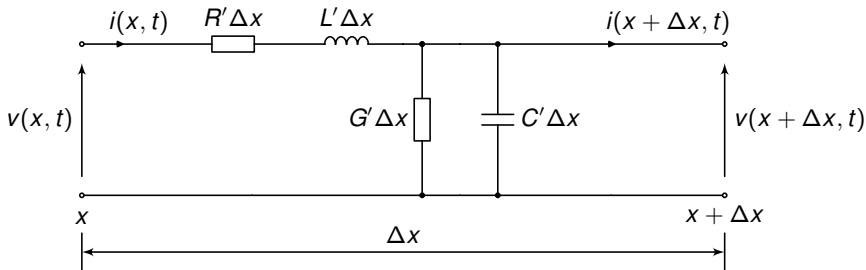
- From KVL

$$v(x + \Delta x, t) = v(x, t) - R' \Delta x i(x, t) - L' \Delta x \frac{\partial i(x, t)}{\partial t}$$

- From KCL

$$i(x + \Delta x, t) = i(x, t) - G' \Delta x v(x + \Delta x, t) - C' \Delta x \frac{\partial v(x + \Delta x, t)}{\partial t}$$

Section of length Δx of homogeneous power line



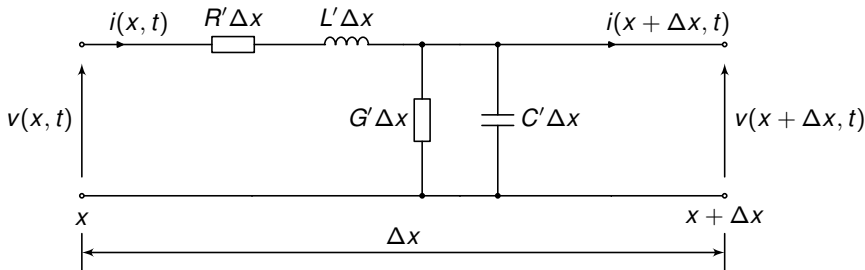
- For infinitesimally small section length $\Delta x \rightarrow 0$, previous equations are equivalent to

$$\frac{\partial v}{\partial x} = - \left(R' + L' \frac{\partial}{\partial t} \right) i$$

$$\frac{\partial i}{\partial x} = - \left(G' + C' \frac{\partial}{\partial t} \right) v$$

5.1 Differential equation of a power line - Telegrapher's equations

Section of length Δx of homogeneous power line



- Decouple equations by differentiating first wrt x and second wrt t and insert resulting expressions in equations (derived by Maxwell around 1860)

$$\frac{\partial^2 v}{\partial x^2} = R' G' v + (R' C' + L' G') \frac{\partial v}{\partial t} + L' C' \frac{\partial^2 v}{\partial t^2}$$
$$\frac{\partial^2 i}{\partial x^2} = R' G' i + (R' C' + L' G') \frac{\partial i}{\partial t} + L' C' \frac{\partial^2 i}{\partial t^2}$$

- In power systems, we are mostly interested in solving telegrapher's equations for special case of *sinusoidal excitation*
- For that case, voltage $v(x, t)$ and current $i(x, t)$ can be represented as phasors with complex amplitudes \underline{V} and \underline{I} and frequency $\omega = 2\pi f$:

$$u(x, t) = \Re \left(\underline{V}(x) e^{j\omega t} \right)$$

$$i(x, t) = \Re \left(\underline{I}(x) e^{j\omega t} \right)$$

- By using phasors, telegrapher's equations reduce to two linear first-order differential equations

$$\frac{d\underline{V}}{dx} = -(R' + j\omega L')\underline{I}$$
$$\frac{d\underline{I}}{dx} = -(G' + j\omega C')\underline{V}$$

- Eliminating $\underline{I}(x)$ leaves us with a linear homogeneous second-order differential equation, which is called *wave equation*

$$\boxed{\frac{d^2 \underline{V}}{dx^2} = (R' + j\omega L')(G' + j\omega C')\underline{V}}$$

- Note: introduction of phasors transforms partial differential equation in ordinary differential equation (i.e. in one variable)

- The solution of the wave equation can be computed as

$$\underline{V}(x) = \underline{V}^+ e^{-\gamma x} + \underline{V}^- e^{\gamma x}$$

- \underline{V}^+ and \underline{V}^- are integration constants
- $\underline{\gamma}$ is called *propagation constant*

$$\underline{\gamma} = \sqrt{(R' + j\omega L')(G' + j\omega C')}$$

- Writing the solution of the wave equation as a function of time, we obtain

$$v(x, t) = \Re \left(\underbrace{V^+ e^{-\gamma x} e^{j\omega t}}_{\text{forward travelling wave}} + \underbrace{V^- e^{\gamma x} e^{j\omega t}}_{\text{backward travelling wave}} \right)$$

- Forward travelling (voltage) wave moves in positive x -direction
- Backward travelling (voltage) wave moves in negative x -direction (also called *reflected wave*)
- Complex propagation constant $\underline{\gamma}$ can be split in real and imaginary part

$$\underline{\gamma} = \alpha + j\beta$$

- α describes damping of (voltage) wave and is measured in Nepers per unit length²
- β describes phase of (voltage) wave at distance x from origin and is measured in radians per unit length

²Np=Neper is a logarithmic unit to measure physical field quantities.

5.2 Solution of the wave equation - Current

- By differentiating $\underline{V}(x)$ we obtain

$$\frac{d\underline{V}}{dx} = -\underline{\gamma}\underline{V}^+ e^{-\underline{\gamma}x} + \underline{\gamma}\underline{V}^- e^{\underline{\gamma}x}$$

and, hence,

$$\begin{aligned} \underline{I}(x) &= \frac{1}{-(R' + j\omega L')} \frac{d\underline{V}}{dx} = \frac{-\underline{\gamma}\underline{V}^+ e^{-\underline{\gamma}x} + \underline{\gamma}\underline{V}^- e^{\underline{\gamma}x}}{-(R' + j\omega L')} \\ &= \sqrt{\frac{G' + j\omega C'}{R' + j\omega L'}} (\underline{V}^+ e^{-\underline{\gamma}x} - \underline{V}^- e^{\underline{\gamma}x}) \end{aligned}$$

- Define *surge impedance* (also called characteristic impedance)

$$\underline{Z}_w = \sqrt{\frac{R' + j\omega L'}{G' + j\omega C'}}$$

$$\Rightarrow \underline{I}(x) = \frac{1}{\underline{Z}_w} (\underline{V}^+ e^{-\underline{\gamma}x} - \underline{V}^- e^{\underline{\gamma}x})$$

- Boundary conditions at beginning of line ($x = 0$)

$$\underline{V}(0) = \underline{V}_1 \quad \underline{I}(0) = \underline{I}_1$$

- Inserting these values in solutions for $\underline{V}(x)$ and $\underline{I}(x)$ at $x = 0$ yields

$$\underline{V}_1 = \underline{V}^+ + \underline{V}^-$$
$$\underline{I}_1 = \frac{\underline{V}^+ - \underline{V}^-}{\underline{Z}_W}$$

- Solving for \underline{V}^+ and \underline{V}^- , we obtain

$$\underline{V}^+ = \frac{\underline{V}_1 + \underline{Z}_W \underline{I}_1}{2}$$
$$\underline{V}^- = \frac{\underline{V}_1 - \underline{Z}_W \underline{I}_1}{2}$$

- Substituting expressions for \underline{V}^+ and \underline{V}^- in equations for $\underline{V}(x)$ and $\underline{I}(x)$ yields

$$\begin{aligned}\underline{V}(x) &= \left(\frac{\underline{V}_1 + \underline{Z}_W \underline{I}_1}{2} \right) e^{-\gamma x} + \left(\frac{\underline{V}_1 - \underline{Z}_W \underline{I}_1}{2} \right) e^{\gamma x} \\ &= \left(\frac{e^{\gamma x} + e^{-\gamma x}}{2} \right) \underline{V}_1 - \underline{Z}_W \underline{I}_1 \left(\frac{e^{\gamma x} - e^{-\gamma x}}{2} \right)\end{aligned}$$

$$\begin{aligned}\underline{I}(x) &= \left(\frac{\underline{V}_1 + \underline{Z}_W \underline{I}_1}{2\underline{Z}_W} \right) e^{-\gamma x} - \left(\frac{\underline{V}_1 - \underline{Z}_W \underline{I}_1}{2\underline{Z}_W} \right) e^{\gamma x} \\ &= -\frac{\underline{V}_1}{\underline{Z}_W} \left(\frac{e^{\gamma x} - e^{-\gamma x}}{2} \right) + \left(\frac{e^{\gamma x} + e^{-\gamma x}}{2} \right) \underline{I}_1\end{aligned}$$

- We can recognise the hyperbolic functions cosh and sinh

$$\cosh(\underline{\gamma}x) = \frac{e^{\underline{\gamma}x} + e^{-\underline{\gamma}x}}{2}, \quad \sinh(\underline{\gamma}x) = \frac{e^{\underline{\gamma}x} - e^{-\underline{\gamma}x}}{2}$$

- Using cosh and sinh gives compact expressions, we obtain equations for propagation of voltage and current from beginning of line

$$\begin{aligned}\underline{V}(x) &= \cosh(\underline{\gamma}x)\underline{V}_1 - \underline{Z}_W \sinh(\underline{\gamma}x)\underline{I}_1 \\ \underline{I}(x) &= -\frac{\underline{V}_1}{\underline{Z}_W} \sinh(\underline{\gamma}x) + \cosh(\underline{\gamma}x)\underline{I}_1\end{aligned}$$

- In same way, we can define boundary conditions at end of line ($x = \ell$)

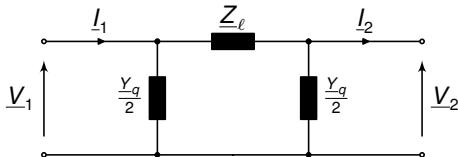
$$\underline{V}(\ell) = \underline{V}_2 \quad \underline{I}(\ell) = \underline{I}_2$$

and obtain equations for propagation of voltage and current from end of line

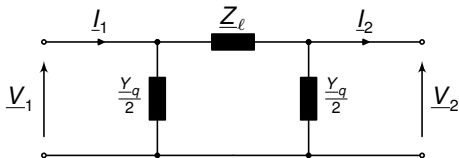
$$\begin{aligned}\underline{V}(x) &= \cosh(\underline{\gamma}(\ell - x))\underline{V}_2 + \underline{Z}_W \sinh(\underline{\gamma}(\ell - x))\underline{I}_2 \\ \underline{I}(x) &= \frac{\underline{V}_2}{\underline{Z}_W} \sinh(\underline{\gamma}(\ell - x)) + \cosh(\underline{\gamma}(\ell - x))\underline{I}_2\end{aligned}$$

- In practice, we often don't need to use the (rather complicated) wave equation to describe phenomena in power systems
 - Reason: Usually, we are interested in the voltage drop across a line or the reactive power flow, but not in the exact trajectory of the voltages and currents along the line
- Then, we may use simplified models for a power line without compromising the accuracy of our calculations too much
- We will discuss such models in the following
 - In particular, we will derive the Π -equivalent circuit of a transmission line

5.3 Π -equivalent circuit of homogeneous power line (1)



- Π -model contains lumped line parameters
- For model derivation, it is convenient to distribute shunt impedance \underline{Y}_q equally on both sides of quadropole
- We will derive this model from the wave equation

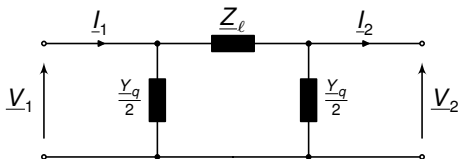


- KCL and KVL yield

$$\begin{bmatrix} \underline{V}_1 \\ \underline{I}_1 \end{bmatrix} = \underbrace{\begin{bmatrix} 1 + \underline{Z}_l \frac{Y_g}{2} & \underline{Z}_l \\ \frac{Y_g}{2} \left(2 + \underline{Z}_l \frac{Y_g}{2} \right) & 1 + \underline{Z}_l \frac{Y_g}{2} \end{bmatrix}}_{=A_1} \begin{bmatrix} \underline{V}_2 \\ \underline{I}_2 \end{bmatrix}$$

- From wave equation we obtain with $\underline{V}_1 = \underline{V}(x=0)$ and $\underline{I}_1 = \underline{I}(x=0)$

$$\begin{bmatrix} \underline{V}_1 \\ \underline{I}_1 \end{bmatrix} = \underbrace{\begin{bmatrix} \cosh(\underline{\gamma}\ell) & \underline{Z}_W \sinh(\underline{\gamma}\ell) \\ \frac{1}{\underline{Z}_W} \sinh(\underline{\gamma}\ell) & \cosh(\underline{\gamma}\ell) \end{bmatrix}}_{=A_2} \begin{bmatrix} \underline{V}_2 \\ \underline{I}_2 \end{bmatrix}$$



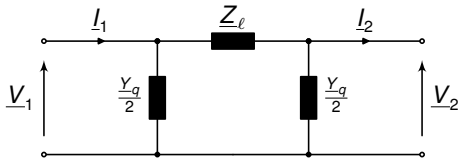
- Comparing coefficients of matrices A_1 and A_2 yields

$$\underline{Z}_\ell = \underline{Z}_W \sinh(\underline{\gamma}\ell)$$

$$\frac{\underline{Y}_q}{2} = \frac{\cosh(\underline{\gamma}\ell) - 1}{\underline{Z}_W \sinh(\underline{\gamma}\ell)} = \frac{1}{\underline{Z}_W} \tanh\left(\frac{\underline{\gamma}\ell}{2}\right)$$

- These parameters correspond to exact relations between currents and voltages according to wave equation for $x = 0$ and $x = \ell$

5.3 Π -equivalent circuit of homogeneous power line - The case $|\underline{\gamma}\ell| \ll 1$



- For $|\underline{\gamma}\ell| \ll 1$, the expressions for \underline{Z}_ℓ and \underline{Y}_q can be simplified

$$\underline{Z}_\ell = \underline{Z}_W \sinh(\underline{\gamma}\ell) \approx \underline{Z}_W \underline{\gamma}\ell = \underline{Z}'\ell$$

$$\frac{\underline{Y}_q}{2} = \frac{1}{\underline{Z}_W} \tanh\left(\frac{(\underline{\gamma}\ell)}{2}\right) \approx \frac{1}{\underline{Z}_W} \frac{\underline{\gamma}\ell}{2} = \frac{\underline{Y}'\ell}{2}$$

→ Concentrated elements \underline{Z}_ℓ and \underline{Y}_q can be computed from distributed parameters R' , L' , G' and C' if $|\underline{\gamma}\ell| \ll 1$

$$\underline{Z}_\ell = \underline{Z}'\ell = (R' + jX')\ell$$
$$\frac{\underline{Y}_q}{2} = \frac{\underline{Y}'\ell}{2} = \frac{(G' + jB')\ell}{2}$$

- Accuracy of assumption $|\underline{\gamma}l| \ll 1$ is crucial for validity of simplified equivalent Π -model
- The larger $|\underline{\gamma}l|$, the worse the model with concentrated parameters \underline{Z}_ℓ and \underline{Y}_q represents evolution of current and voltage along the line
- Whenever you use a simplified Π -model to represent a power line, be aware that the model accuracy reduces with increasing line length!
- Rule of thumb:
 - Max. length for overhead line ≈ 300 km
 - Max. length for cable ≈ 100 km
- Therefore, long power lines are often split into several shorter sections in power flow calculations and each section is represented by individual (simplified) Π -model

Task. Consider a power line with the following characteristics

$$R' = 0.05 \Omega/\text{km}, \quad L' = 1.25 \text{ mH}/\text{km}, \quad G' = 0 \mu\text{ S}/\text{km}, \quad C' = 10 \text{ nF}/\text{km}.$$

Suppose that the line length is 200 km and that the line is operated with a frequency of 50 Hz.

- 1) Calculate the series impedance \underline{Z}_ℓ^E and the shunt admittance \underline{Y}_q^E for the exact Π -equivalent circuit.
- 2) If $|\underline{\gamma}\ell| \ll 1$, then calculate the simplified series impedance \underline{Z}_ℓ and the shunt admittance \underline{Y}_q of the Π -equivalent circuit for that case.

Solution. 1) Surge impedance of line with $G' = 0$ and $\omega = 2\pi f = 314.16$ rad/s

$$\begin{aligned} Z_w &= \sqrt{\frac{R' + j\omega L'}{j\omega C'}} \\ &= \sqrt{\frac{0.05 + j314.16 \cdot 1.25 \cdot 10^{-3}}{j314.16 \cdot 10 \cdot 10^{-9}}} = 354.27 - j22.463 = 354.98 \angle -3.63^\circ \Omega \end{aligned}$$

Propagation constant of line with $G' = 0$ and $\omega = 2\pi f = 314.16$ rad/s

$$\begin{aligned} \gamma &= \sqrt{(R' + j\omega L')(j\omega C')} \\ &= \sqrt{(0.05 + j314.16 \cdot 1.25 \cdot 10^{-3})(j314.16 \cdot 10 \cdot 10^{-9})} \\ &= 0.0001 \text{ Np/km} + j0.0011 \text{ [rad/km]} \end{aligned}$$

Np=Neper (logarithmic unit to measure physical field quantities)

Series impedance of Π -equivalent circuit

$$\begin{aligned}\underline{Z}_\ell^E &= \underline{Z}_W \sinh(\gamma \ell) \\ &= 354.98 \angle -3.63^\circ \cdot \sinh(0.0011 \angle 84.81^\circ \cdot 200) \\ &= 11.818 + j76.889 = 77.79 \angle 81.27^\circ \Omega\end{aligned}$$

Shunt admittance of Π -equivalent circuit

$$\begin{aligned}\underline{Y}_q^E &= \frac{2}{\underline{Z}_W} \tanh\left(\frac{\gamma \ell}{2}\right) \\ &= \frac{2}{354.98 \angle -3.63^\circ} \cdot \tanh\left(\frac{0.0011 \angle 84.81^\circ \cdot 200}{2}\right) \\ &= 1.754 \cdot 10^{-5} + j6.246 \cdot 10^{-4} = 6.248 \cdot 10^{-4} \angle 88.40^\circ \text{ S}\end{aligned}$$

→ Exact Π -equivalent circuit can have shunt conductance even if³ $G' = 0!$

³Physical explanation: We could model the considered line equivalently by two Π -equivalent circuits in series. Then, we would see that there are active power losses in the circuit. Thus, the single Π -equivalent circuit has to have an ohmic component.

2) We have that

$$|\underline{\gamma}l| = 0.0011 \cdot 200 = 0.22$$

This value is reasonably smaller than 1

Thus, using the simplified equations valid for $|\underline{\gamma}l| \ll 1$, we obtain

$$\begin{aligned}\underline{Z}_\ell &= (R' + j\omega L')\ell \\ &= (0.05 + j314.16 \cdot 1.25 \cdot 10^{-3}) \cdot 200 \\ &= 10 + j78.54 = 79.17 \angle 82.75^\circ \Omega\end{aligned}$$

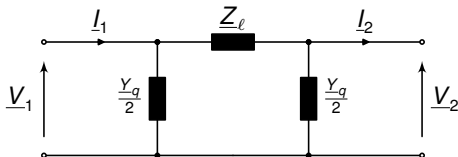
and

$$\begin{aligned}\underline{Y}_q &= jB'\ell = j\omega C'\ell \\ &= j314.16 \cdot 10 \cdot 10^{-9} \cdot 200 = j6.283 \cdot 10^{-4} \text{ S}\end{aligned}$$

→ Simplified Π -equivalent circuit has NO shunt conductance whenever $G' = 0!$

Remaining parameters are very similar to exact values:

$$\underline{Z}_\ell \approx \underline{Z}_\ell^E, \mathfrak{S}(\underline{Y}_q) \approx \mathfrak{S}(\underline{Y}_q^E)$$



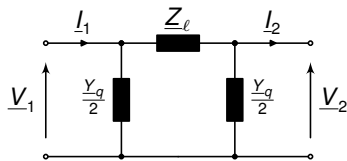
- In practice, G' is small (in particular for voltages $> 110\text{kV}$) and therefore often neglected
- Then, shunt admittance is purely capacitive

$$\underline{Z}_l = \underline{Z}'l = (R' + jX')l \qquad \frac{Y_g}{2} = \frac{Y'}{2}l = \frac{jB'}{2}l$$

- Some times, also conductor resistances neglected $\rightarrow R' = 0$; such line model is called *lossless* and its concentrated (or lumped) parameters are given by

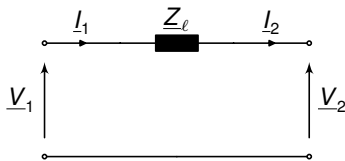
$$\underline{Z}_l = \underline{Z}'l = jX'l \qquad \frac{Y_g}{2} = \frac{Y'}{2}l = \frac{jB'}{2}l = \frac{j\omega C'}{2}l$$

- For *overhead lines* models can be further simplified
- Typically, overhead lines classified into 3 categories
 - **Short lines (up to 100 km).** Usually, C' and G' very small; model: series impedance $\underline{Z}_\ell = R'\ell + j\omega L'\ell$; shunt admittance \underline{Y}_q is completely neglected
 - **Medium length lines (100 to 300 km).** Use of simplified Π -model with $G' = 0$ without any significant loss of accuracy
 - **Long lines (larger than 300 km).** Significant inaccuracies with concentrated parameter model. Line should either be represented by wave equation or split into several shorter sections



Medium length line model

$$\underline{Y}_q = jB'\ell$$



Short line model

$$\underline{Y}_q = 0$$

- We compare results obtained with different models for exemplary 230 kV transmission line with characteristic impedance and propagation constant

$$\underline{Z}_w = 382.2 - j16.5 \Omega \quad \underline{\gamma} = \alpha + j\beta = 0.0001 \text{ [Np/km]} + j0.0011 \text{ [rad/km]}$$

Np=Neper (logarithmic unit to measure physical field quantities)

- We seek to calculate voltage \underline{V}_2 at end of line under *no load* conditions
 $\rightarrow I_2 = 0$
- We assume $|\underline{V}_1| = 1.0 \text{ pu}$
- We will explore 3 different ways
 - 1) Using the exact wave equation (Section 4.2)
 - 2) Using the medium length Π -equivalent circuit (Section 4.3)
 - 3) Using the short line model (Section 4.4)

- 1) For $I_2 = 0$, exact wave equation reduces to (see matrix A_2)

$$\underline{V}_1 = \underline{V}(x = 0) = \underline{V}_2 \cosh(\underline{\gamma}\ell)$$

- 2) Medium length Π -model

$$\underline{Z}_\ell \approx \underline{Z}_W \underline{\gamma}\ell \quad \frac{Y_q}{2} \approx \frac{1}{\underline{Z}_W} \frac{\underline{\gamma}\ell}{2}$$

Hence,

$$\underline{V}_1 = \left(1 + \frac{\underline{Z}_\ell Y_q}{2}\right) \underline{V}_2 = \left(1 + \frac{(\underline{\gamma}\ell)^2}{2}\right) \underline{V}_2$$

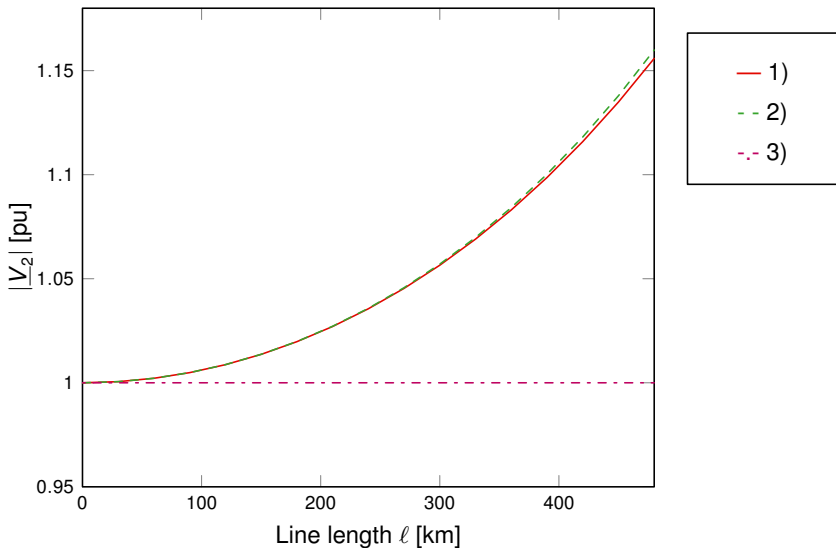
- 3) Short line model: there is no voltage drop across series element for zero current $\rightarrow \underline{V}_2 = \underline{V}_1$

Values for $|V_2|$ obtained for different line lengths and different models

Length ℓ in km	$ \underline{\gamma}\ell $	1)	2)	3)
50	0.0552	1.0015	1.0015	1.000
100	0.1105	1.0060	1.0060	1.000
300	0.3314	1.0565	1.0570	1.000
500	0.5523	1.1710	1.1759	1.000

- For short line lengths (< 50 km) all models provide almost identical results
- With increasing line length, results with short line clearly differ from those with other models
- Accuracy of Π -model fairly good up to 300 km, but increasing deviation with increasing length

5.4 Model simplifications - Comparison: Plots



- Overhead lines most economic solution for long-distance power transmission
- An overhead line consists of conductors, support structures, shield wires and insulators
- Characteristics of power lines can be represented by set of concentrated parameters R' , L' , C' and G'
- Exact propagation of voltage and current in a power line can be described by telegrapher's equations (in time-domain), respectively by the wave equation (in phasor-domain)
- For most practical applications, the use of a Π -equivalent circuit suffices to accurately describe the voltage and current relations on a power line
- The validity of the Π -model reduces significantly for long lines (>300 km)